

# Unidad 5: Inductancia

Circuitos de corriente alterna

# Inductancia

Ley de Faraday

$$\varepsilon = -\frac{d\phi_B}{dt}$$

$$\varepsilon_L = -L \frac{di}{dt}$$

$L$ : inductancia

$$\varepsilon_L = -N \frac{d\phi_B}{dt}$$



$$L = \frac{N\phi_B}{i}$$

$$L = \frac{\varepsilon_L}{\frac{di}{dt}}$$

$$R = \frac{\Delta V}{I}$$

La inductancia presenta un efecto similar al de la resistencia

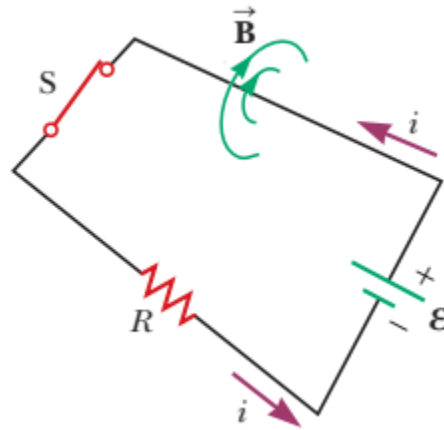
Unidades para  $L \rightarrow [Hy]; \left[ \frac{V \cdot s}{A} \right]$

Para el caso de una bobina

$$\phi_B = B \cdot A = \mu_0 n I A = \mu_0 \frac{N}{l} I A$$



$$L = \frac{N\phi_B}{i} = \frac{\mu_0 N^2 A}{l}$$



f.e.m autoinducida

Autoinducción

# Caso

Si consideramos una bobina de 280 vueltas, 290 mm de longitud y una sección de 5 cm<sup>2</sup>. Calcular la inductancia y la f.e.m autoinducida si la corriente disminuye a 40 A/s.

$$L = \frac{\mu_0 N^2 A}{l} = 4\pi \times 10^{-7} \left[ \frac{T \cdot m}{A} \right] \frac{280^2}{0,290[m]} 5 \times 10^{-4} [m^2] = 0,1698 [mH]$$

$$\varepsilon_L = -L \frac{di}{dt} = -0,0001698 [H] \cdot (-40) \left[ \frac{A}{s} \right] = 6,792 [mV]$$

# Autoinductancia



# Circuitos RL

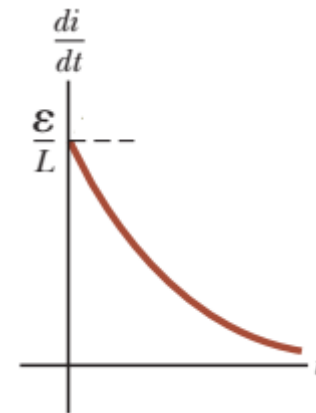
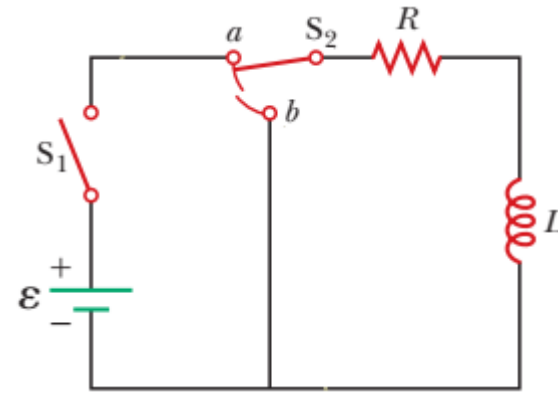
$$\varepsilon - iR - L \frac{di}{dt} = 0$$

$$\frac{di}{dt} + i \frac{R}{L} = \frac{\varepsilon}{L} \quad \Rightarrow \quad \frac{di}{dt} = \frac{\varepsilon}{L} - i \frac{R}{L}$$

$$\frac{di}{dt} = \frac{R}{L} \left( \frac{\varepsilon}{R} - i \right) \quad \Rightarrow \quad \frac{di}{\left( \frac{\varepsilon}{R} - i \right)} = \frac{R}{L} dt \quad \Rightarrow \quad d \left( \frac{\varepsilon}{R} - i \right) = -di$$

$$\frac{d \left( \frac{\varepsilon}{R} - i \right)}{\left( \frac{\varepsilon}{R} - i \right)} = -\frac{R}{L} dt \quad \Rightarrow \quad \ln \left( \frac{\varepsilon}{R} - i \right) = -\frac{R}{L} t + \ln K \quad \Rightarrow \quad \frac{\varepsilon}{R} - i = K e^{-\frac{R}{L} t} \quad \Rightarrow \quad i = \frac{\varepsilon}{R} - K e^{-\frac{R}{L} t}$$

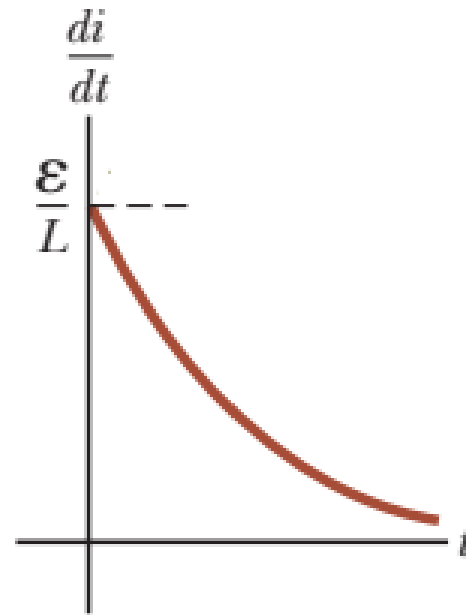
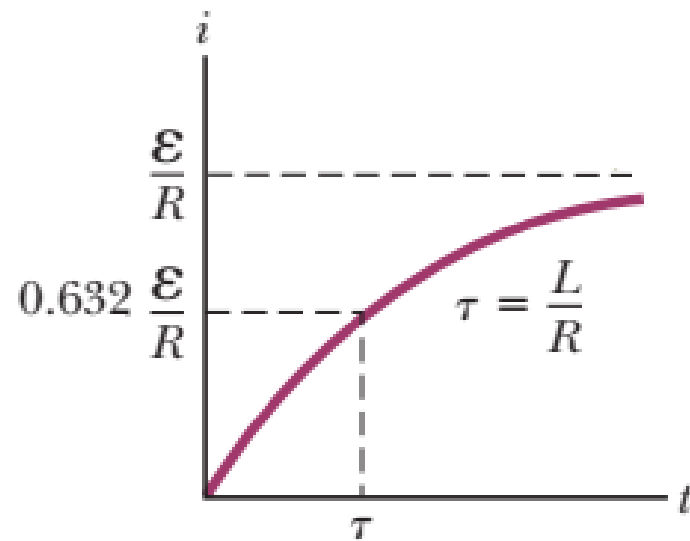
$$\text{Para } t = 0 \rightarrow i = 0 \quad K = \frac{\varepsilon}{R} \quad \Rightarrow \quad i = \frac{\varepsilon}{R} \left( 1 - e^{-\frac{R}{L} t} \right)$$



# Circuitos RL

Constante de Tiempo

$$\tau = \frac{L}{R} [s]$$



# Caso circuito RL

Consideremos un circuito RL donde la fuente tiene  $\varepsilon = 12[V]$ ,  $R = 5[\Omega]$ ,  $L = 40[mH]$

$$i = \frac{\varepsilon}{R} \left(1 - e^{-\frac{R}{L}t}\right) = \frac{12[V]}{5[\Omega]} \left(1 - e^{-\frac{5[\Omega]}{40 \times 10^{-3}[H]}t}\right) = 2,4(1 - e^{-125t})[A]$$

$$t = 2[ms] \quad i = 2,4(1 - e^{-125t})[A] = 0,53[A]$$

$$\tau = \frac{L}{R} = \frac{40 \times 10^{-3}[H]}{5[\Omega]} = 0,008[s]$$

# Energía de un campo magnético

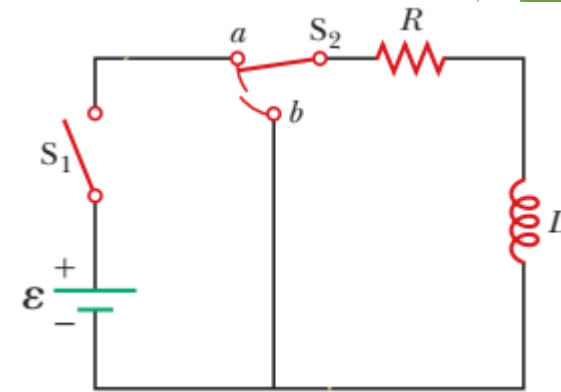
$$\varepsilon = iR + L \frac{di}{dt} \quad i\varepsilon = i^2R + Li \frac{di}{dt} \quad \frac{dU_B}{dt} = Li \frac{di}{dt}$$

$$U_B = \int dU_B = \int_0^i Li \, di = L \int_0^i i \, di$$

$$U_B = \frac{1}{2} Li^2 \quad L = \mu_0 n^2 V \quad B = \mu_0 ni \quad i = \frac{B}{\mu_0 n}$$

$$U_B = \frac{1}{2} \mu_0 n^2 V \left( \frac{B}{\mu_0 n} \right)^2 = \frac{B^2}{2\mu_0} V$$

$$u_B = \frac{U_B}{V} \quad u_B = \frac{B^2}{2\mu_0}$$



*V: volumen de la bobina*



# Inductancia mutua

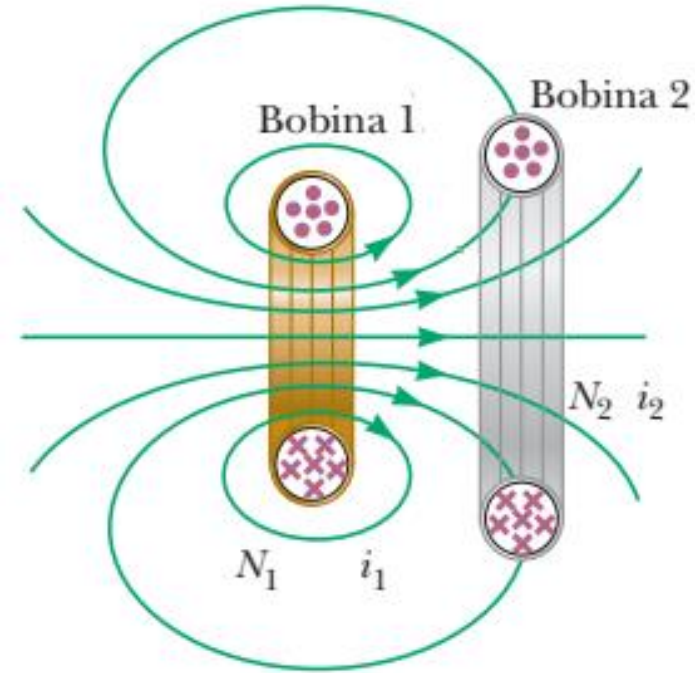
$$\varepsilon = -N \frac{d\phi_B}{dt} \quad L = \frac{N\phi_B}{i}$$

$$M_{12} = \frac{N_2\phi_{12}}{i_1}$$

$$\varepsilon_2 = -N_2 \frac{d\phi_{12}}{dt} = -N_2 \frac{d}{dt} \left( \frac{M_{12}i_1}{N_2} \right) = -M_{12} \frac{di_1}{dt}$$

$$\varepsilon_1 = -M_{21} \frac{di_2}{dt} \quad M_{12} = M_{21}$$

$$\varepsilon_1 = -M \frac{di_2}{dt} \quad \varepsilon_2 = -M \frac{di_1}{dt}$$

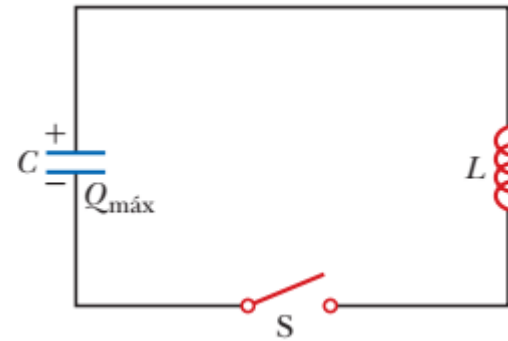


# Oscilaciones en un circuito LC

Previo al cierre del interruptor, el capacitor se encuentra cargado.

$$t < 0 \quad U_C = \frac{Q_{\max}^2}{2C} \quad U_B = 0$$

$$t = 0 \quad U_C = \frac{Q^2}{2C} \rightarrow 0 \quad U_B = \frac{1}{2} Li^2$$



# Oscilaciones en un circuito LC

Energía almacenada en un circuito LC

$$U = U_C + U_B = \frac{q^2}{2C} + \frac{1}{2}Li^2 = \frac{Q_{max}^2}{2C} \quad \longrightarrow \quad \frac{dU}{dt} = \frac{d}{dt} \left( \frac{q^2}{2C} + \frac{1}{2}Li^2 \right) = \frac{q}{C} \frac{dq}{dt} + Li \frac{di}{dt} = 0$$

$$i = \frac{dq}{dt} \quad \longrightarrow \quad \frac{di}{dt} = \frac{d^2q}{dt^2} \quad \longrightarrow \quad \frac{q}{C} \frac{dq}{dt} + Li \frac{d^2q}{dt^2} = 0 \quad \longrightarrow \quad \frac{d^2q}{dt^2} = -\frac{1}{LC}q$$

$$q = Q_{max} \cos(\omega t + \varphi) \quad \omega = \frac{1}{\sqrt{LC}}$$

$$i = \frac{dq}{dt} = -\omega Q_{max} \sin(\omega t + \varphi)$$

Para determinar el valor  $\varphi$  consideramos las condiciones iniciales, donde para  $t = 0 \rightarrow i = 0$

Por lo tanto:  $\varphi = 0$ .

# Oscilaciones en un circuito LC

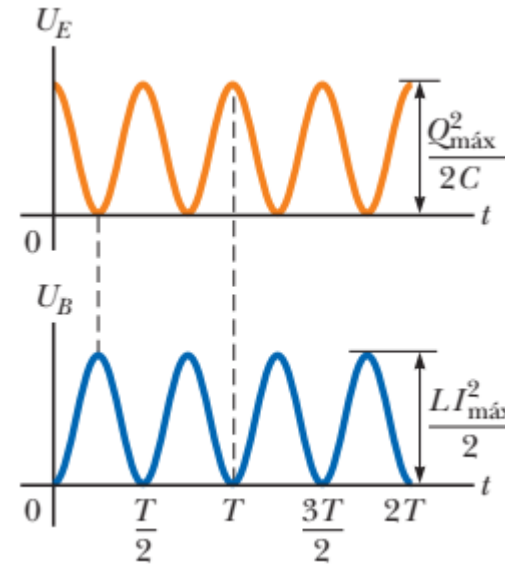
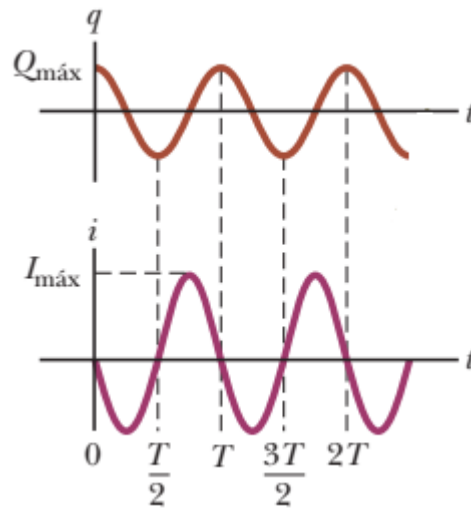
$$q = Q_{max} \cos \omega t$$

$$i = -\omega Q_{max} \sin \omega t = -I_{max} \sin \omega t$$

$$U = U_C + U_B = \frac{Q_{max}^2}{2C} \cos^2 \omega t + \frac{1}{2} L I_{max}^2 \sin^2 \omega t$$

$$\cos^2 \omega t + \sin^2 \omega t = 1$$

$$\frac{Q_{max}^2}{2C} = \frac{1}{2} L I_{max}^2$$



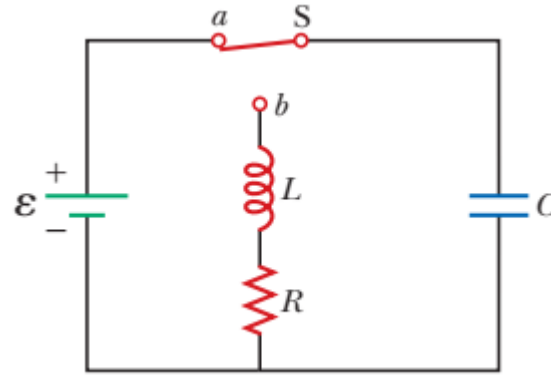
# Circuito RLC

$$\frac{dU}{dt} = -i^2 R \quad U = U_C + U_B$$

$$\frac{q}{C} \frac{dq}{dt} + Li \frac{di}{dt} + i^2 R = 0 \quad i = \frac{dq}{dt}$$

$$Li \frac{d^2 q}{dt^2} + i^2 R + \frac{q}{C} i = 0$$

$$\frac{d^2 q}{dt^2} + \frac{R}{L} \frac{dq}{dt} + \frac{q}{LC} = 0$$



→ Ecuación Diferencial, 2° Orden, homogénea

$$D^2 + \frac{R}{L} D + \frac{1}{LC} = 0$$

→ Ecuación característica

# Circuito RLC

$$D^2 + \frac{R}{L}D + \frac{1}{LC} = 0$$

$$D_1 = \frac{-\frac{R}{L} + \sqrt{\left(\frac{R}{L}\right)^2 - \frac{4}{LC}}}{2}$$

$$D_1 = \alpha + \beta$$

$$D_2 = \frac{-\frac{R}{L} - \sqrt{\left(\frac{R}{L}\right)^2 - \frac{4}{LC}}}{2}$$

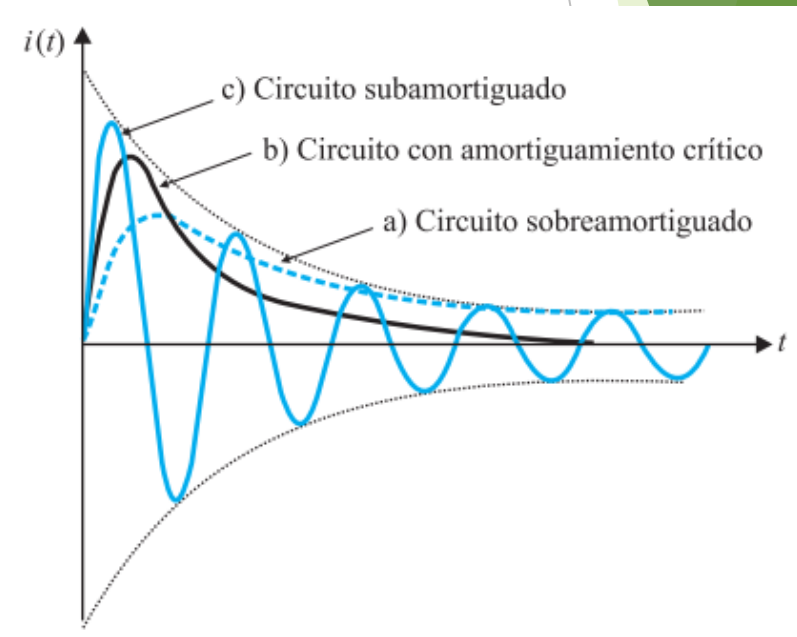
$$D_2 = \alpha - \beta$$

# Circuito RLC

Si  $\left(\frac{R}{L}\right)^2 > \frac{4}{LC}$ , las raíces de  $D_1$  y  $D_2$  son reales y distintas  $\rightarrow$  Oscilación sobreamortiguada.

Si  $\left(\frac{R}{L}\right)^2 = \frac{4}{LC}$ , las raíces de  $D_1$  y  $D_2$  son reales y distintas  $\rightarrow$  Oscilación crítica.

Si  $\left(\frac{R}{L}\right)^2 < \frac{4}{LC}$ , las raíces de  $D_1$  y  $D_2$  no son reales y distintas  $\rightarrow$  Oscilación subamortiguada

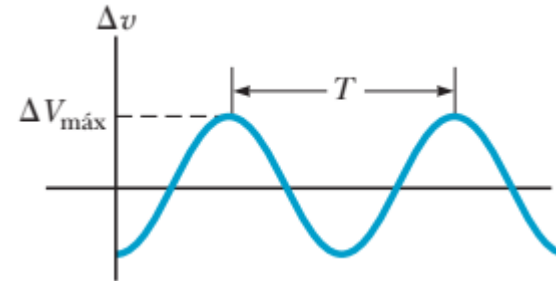


# Fuentes de corriente alterna

$$\Delta v = \Delta V_{max} \sin(\omega t + \varphi)$$

$\Delta V_{max}$ : amplitud máxima de la onda senoidal

$$\omega = 2\pi f = \frac{2\pi}{T} \quad f = \frac{1}{T}$$





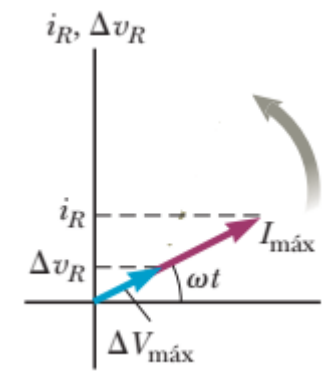
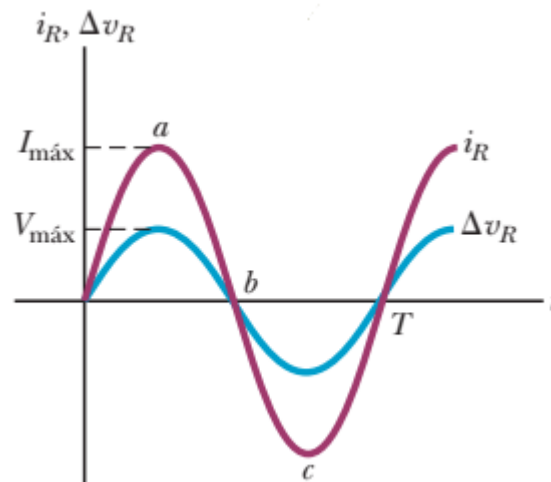
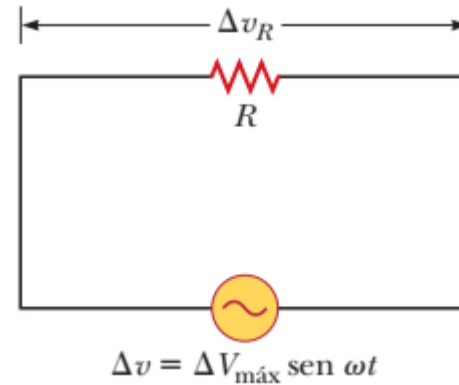
# Resistores en un circuito de CA

$$\Delta v - iR = 0$$

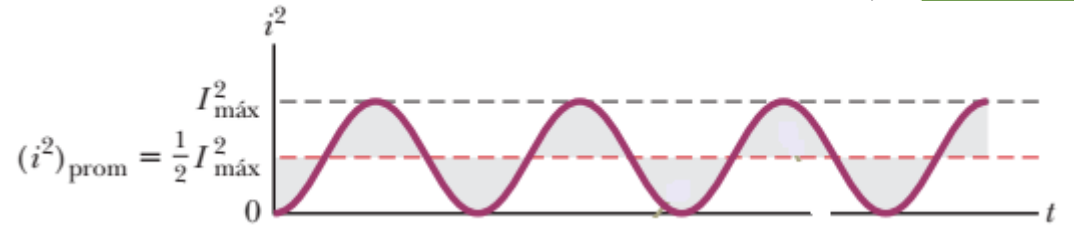
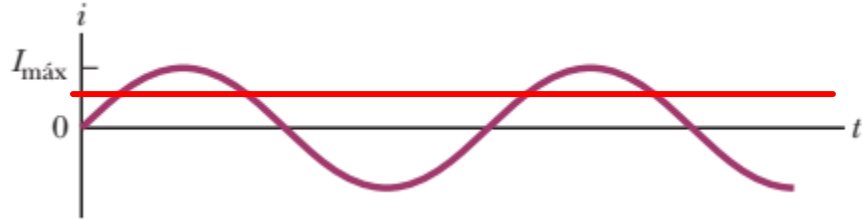
$$i = \frac{\Delta v}{R} = \frac{\Delta V_{\text{máx}}}{R} \sin(\omega t + \varphi) \quad \varphi = 0^\circ$$

$$I_{\text{max}} = \frac{\Delta V_{\text{máx}}}{R}$$

$$\Delta v_R = i_R R = I_{\text{máx}} R \sin \omega t$$



# Resistores en un circuito de CA



$$P = i^2 \cdot R$$

$$I_{ef}^2 = \frac{1}{T} \int i^2 dt \quad \rightarrow \quad I_{ef} = \sqrt{\frac{1}{T} \int i^2 dt} \quad \rightarrow \quad I_{ef} = \frac{I_{max}}{\sqrt{2}} = 0,707 I_{max}$$

# Inductores en un circuito de CA

$$\Delta v - \Delta v_L = 0 \quad \Delta v - L \frac{di_L}{dt} = 0$$

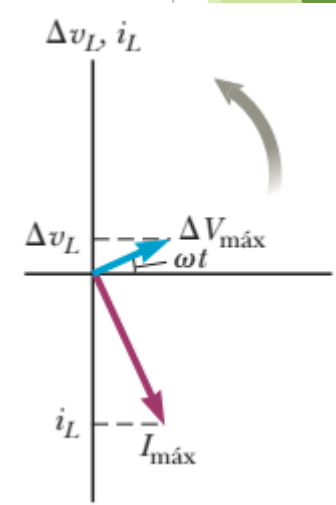
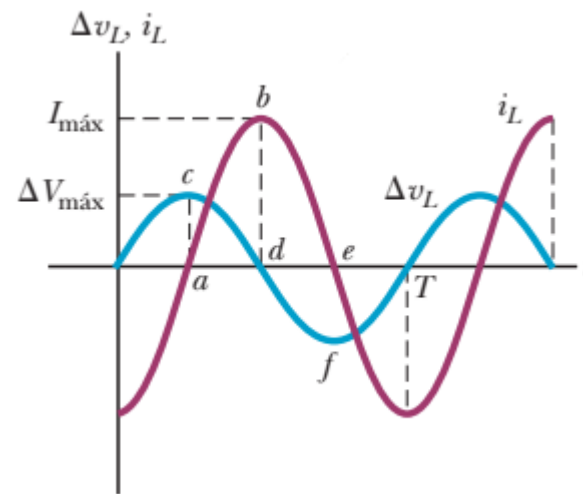
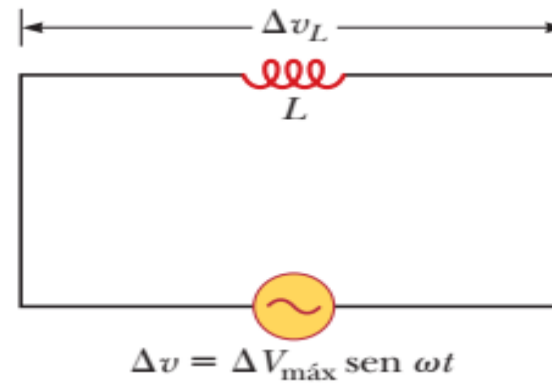
$$\Delta v = L \frac{di_L}{dt} = \Delta V_{max} \sin \omega t$$

$$di_L = \frac{\Delta V_{max}}{L} \Delta V_{max} \sin \omega t dt$$

$$i_L = \frac{\Delta V_{max}}{L} \int \sin \omega t dt = -\frac{\Delta V_{max}}{\omega L} \cos \omega t$$

$$\cos \omega t = -\sin \left( \omega t - \frac{\pi}{2} \right)$$

$$i_L = \frac{\Delta V_{max}}{\omega L} \sin \left( \omega t - \frac{\pi}{2} \right)$$



# Inductores en un circuito de CA

$$I_{max} = \frac{\Delta V_{max}}{\omega L}$$

$$X_L = \omega L = 2\pi fL$$

Reactancia inductiva

$$I_{max} = \frac{\Delta V_{max}}{X_L}$$

$$\Delta v_L = -L \frac{di_L}{dt} = \Delta V_{mac} \sin \omega t = -I_{max} X_L \sin \omega t$$

# Capacitores en un circuito de CA

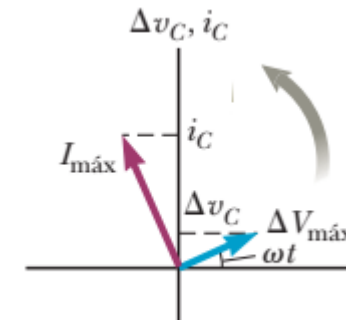
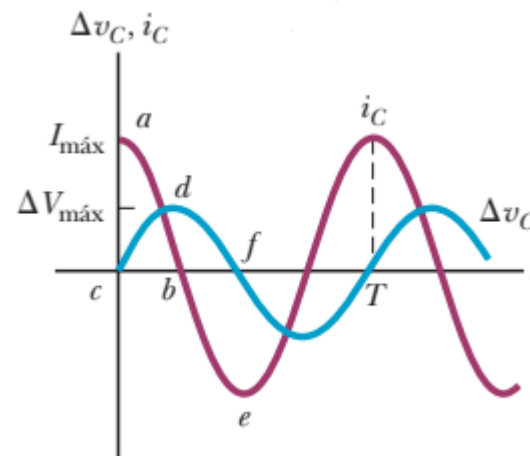
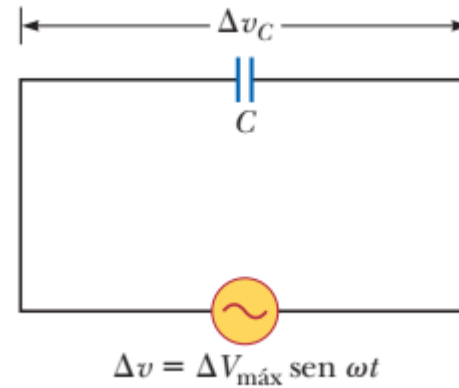
$$\Delta v - \Delta v_C = 0 \quad \Delta v - \frac{q}{C} = 0$$

$$q = C \Delta V_{max} \sin \omega t$$

$$i_C = \frac{dq}{dt} = \omega C \Delta V_{max} \cos \omega t$$

$$\cos \omega t = \sin \left( \omega t + \frac{\pi}{2} \right)$$

$$i_C = \omega C \Delta V_{max} \sin \left( \omega t + \frac{\pi}{2} \right)$$



# Capacitores en un circuito de CA

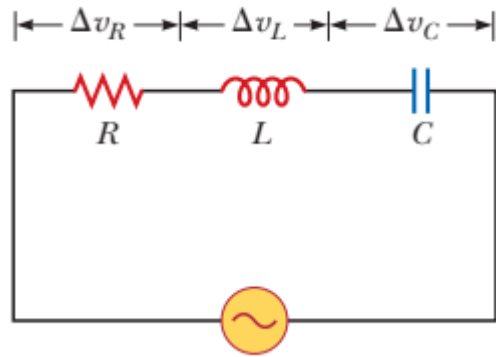
$$I_{max} = \omega C \Delta V_{max} = \frac{\Delta V_{max}}{\frac{1}{\omega C}}$$

$$X_C = \frac{1}{\omega C} = \frac{1}{2\pi f C} \quad \text{Reactancia capacitiva}$$

$$I_{max} = \frac{\Delta V_{max}}{X_C}$$

$$\Delta v_C = \Delta V_{max} \sin \omega t = I_{max} X_C \sin \omega t$$

# Circuito RLC en serie

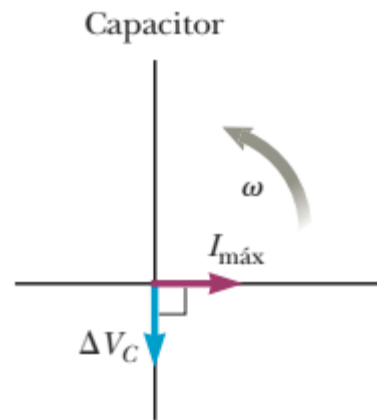
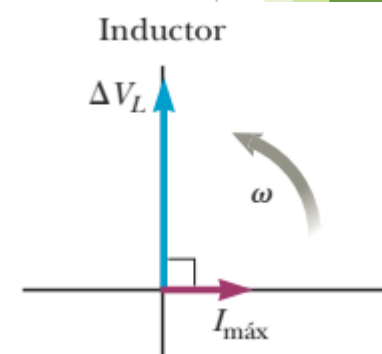
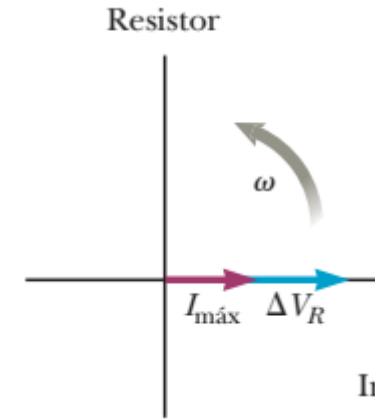
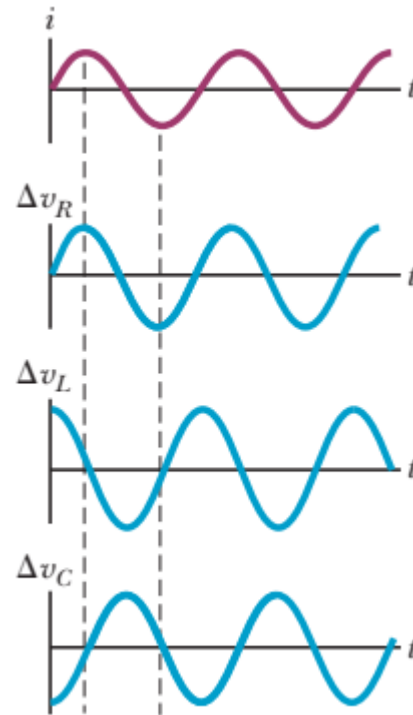


$$i = I_{max} \sin(\omega t + \phi)$$

$$\Delta v_R = I_{max} R \sin \omega t = \Delta V_R \sin \omega t$$

$$\Delta v_L = I_{max} X_L \sin\left(\omega t + \frac{\pi}{2}\right) = \Delta V_L \cos \omega t$$

$$\Delta v_C = I_{max} X_C \sin\left(\omega t - \frac{\pi}{2}\right) = -\Delta V_C \cos \omega t$$



# Circuito RLC en serie

$$\Delta V_{max} = \sqrt{\Delta V_R^2 + (\Delta V_L - \Delta V_C)^2} = \sqrt{(I_{max}R)^2 + (I_{max}X_L - I_{max}X_C)^2} = I_{max}\sqrt{R^2 + (X_L - X_C)^2}$$

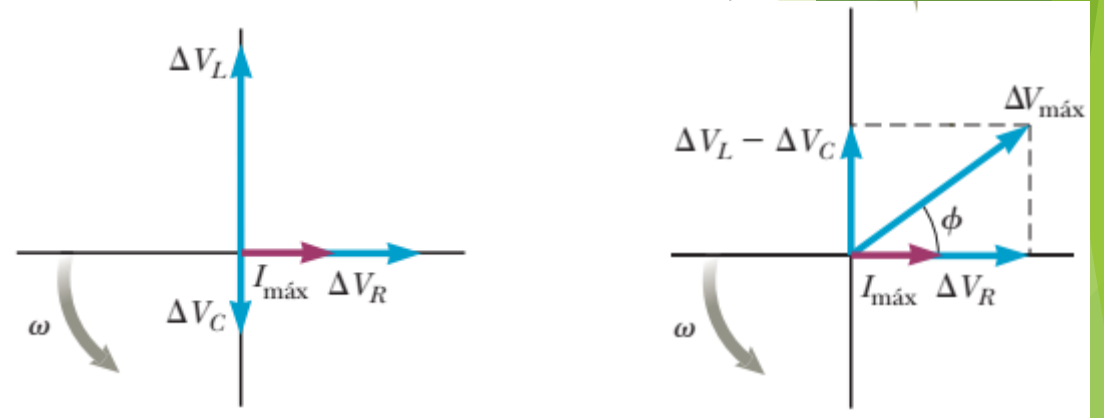
$$I_{max} = \frac{\Delta V_{max}}{\sqrt{R^2 + (X_L - X_C)^2}}$$

$$Z = \sqrt{R^2 + (X_L - X_C)^2} \quad \text{Impedancia}$$

$$I_{max} = \frac{\Delta V_{max}}{Z}$$

$$\phi = \text{arc tg} \left( \frac{\Delta V_L - \Delta V_C}{\Delta V_R} \right) = \text{arc tg} \left( \frac{I_{max}X_L - I_{max}X_C}{I_{max}R} \right)$$

$$\phi = \text{arc tg} \left( \frac{X_L - X_C}{R} \right)$$





# Circuito RLC en serie

$$i = \sqrt{2} I \sin(\omega t + \phi) \quad I_{ef} = \frac{I_{m\acute{a}x}}{\sqrt{2}}$$

$$\Delta v = \sqrt{2} \Delta V \sin \omega t$$

$$p = \Delta V \cdot i = \sqrt{2} \Delta V I \sin \omega t \sqrt{2} I \sin(\omega t + \phi) = 2\Delta V I \sin(\omega t + \phi) \sin \omega t$$

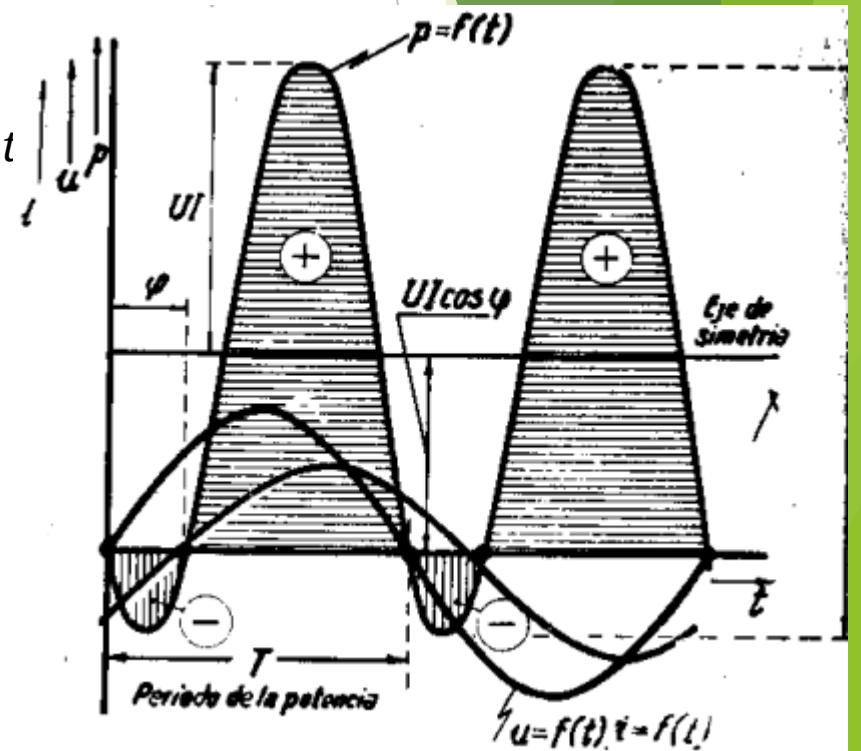
$$= 2\Delta V I [\sin^2 \omega t \cos \phi + \sin \omega t \cos \omega t \sin \phi]$$

$$\sin^2 \omega t = \frac{1 - \cos 2\omega t}{2} \quad \sin \omega t \cos \omega t = \frac{\sin 2\omega t}{2}$$

$$p = \Delta V I [\cos \phi - \cos 2\omega t \cos \phi + \sin 2\omega t \sin \phi]$$

$$p = \Delta V I \cos \phi - \Delta V I \cos (2\omega t \pm \phi)$$

$$P = \frac{1}{T} \int_0^T p \, dt = \frac{1}{T} \Delta V I \int_0^T \cos \phi - \cos (2\omega t \pm \phi) dt = \Delta V I \cos \phi \rightarrow \text{Potencia activa}$$



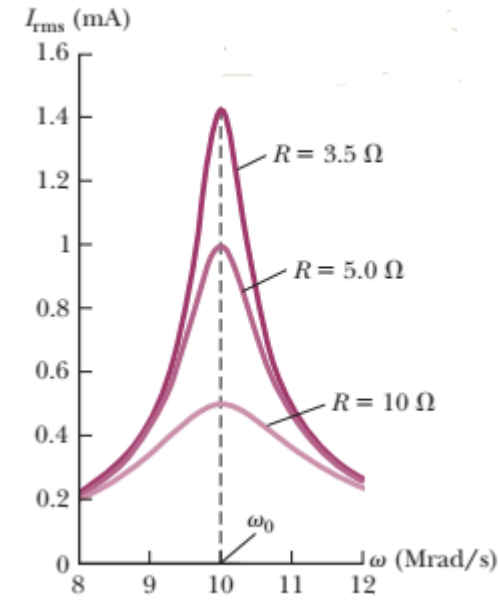
# Resonancia en circuito RLC serie

$$I_{m\acute{a}x} = \frac{\Delta V_{ef}}{Z}$$

$$I_{ef} = \frac{\Delta V_{ef}}{\sqrt{R^2 + (X_L - X_C)^2}}$$

$$X_L - X_C = 0 \quad \text{Frecuencia de resonancia}$$

$$\omega_0 = \frac{1}{\sqrt{LC}}$$



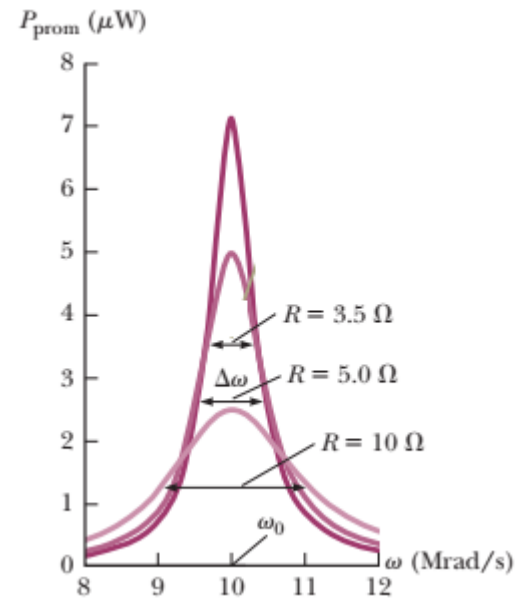
# Resonancia en circuito RLC serie

$$P_{promedio} = \frac{(\Delta V_{ef})^2 R \omega^2}{R^2 \omega^2 + L^2 (\omega^2 - \omega_0^2)^2}$$

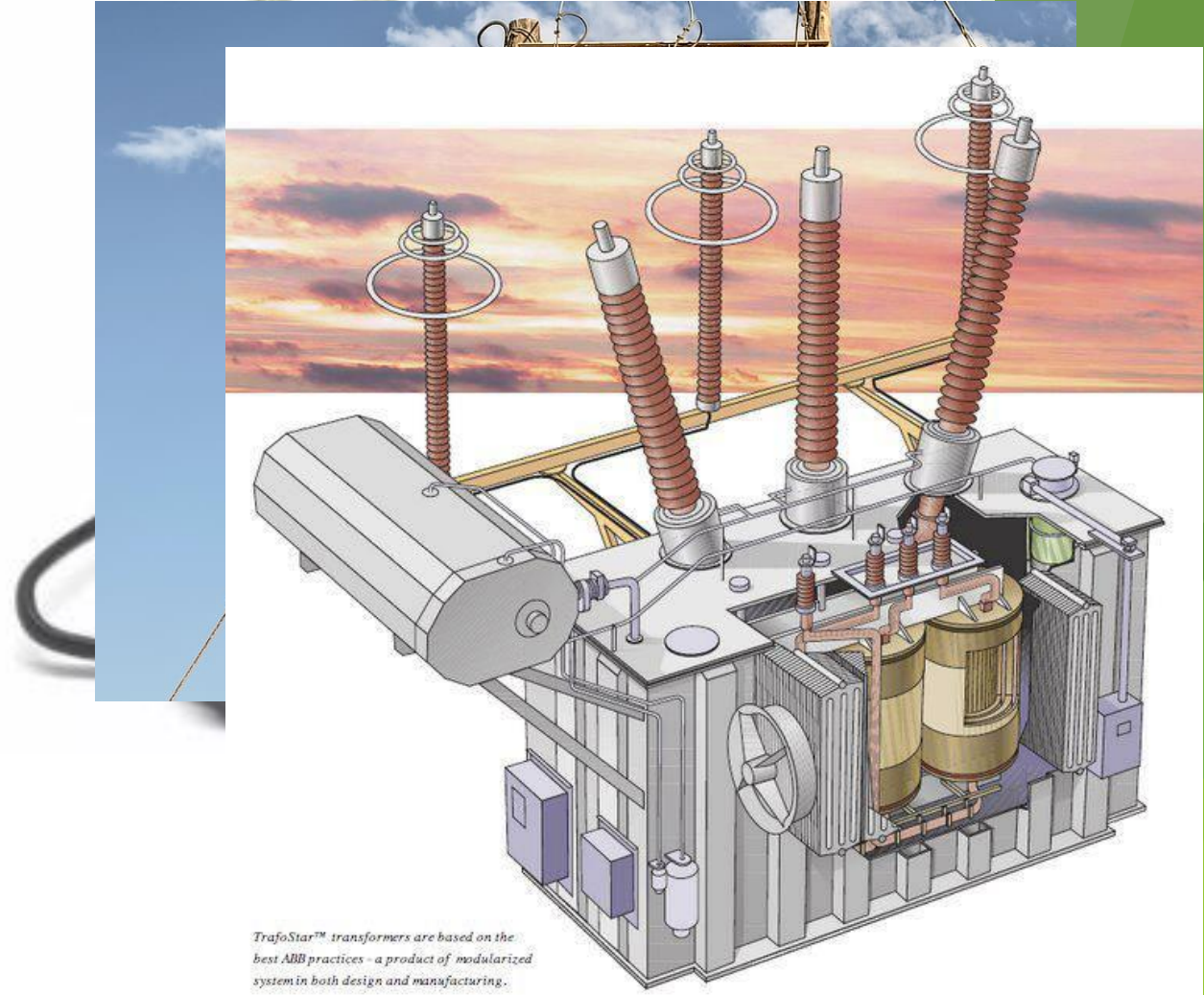
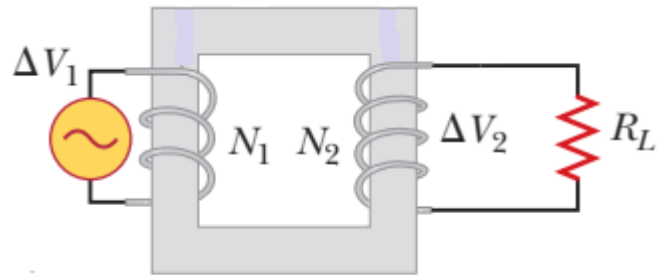
$$Q = \frac{\omega_0}{\Delta\omega} \quad \text{Factor de calidad}$$

$$\Delta\omega = \frac{R}{L}$$

$$Q = \frac{\omega_0 L}{R}$$



# Transformación y transmisión de potencia



*TrafoStar™ transformers are based on the best ABB practices - a product of modularized system in both design and manufacturing.*

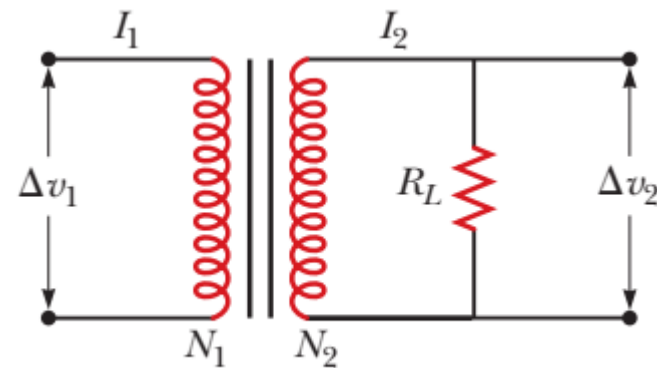
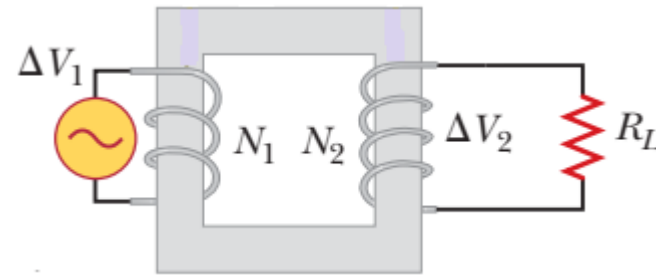
# Transformación y transmisión de potencia

$$\Delta V_1 = -N_1 \frac{d\phi_B}{dt}$$

$$\Delta V_2 = -N_2 \frac{d\phi_B}{dt}$$

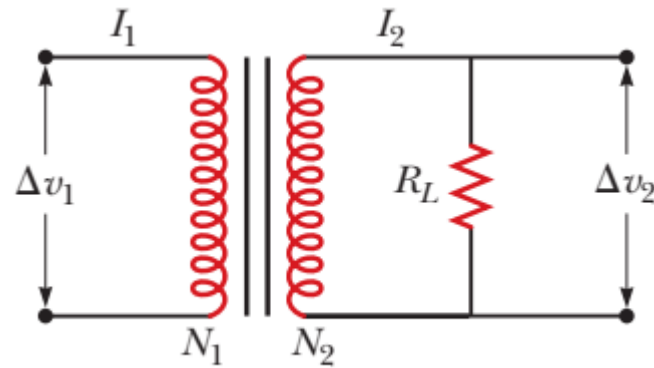
$$\frac{\Delta V_1}{\Delta V_2} = \frac{N_1}{N_2} \quad \rightarrow \quad \Delta V_1 = \frac{N_1}{N_2} \Delta V_2$$

$$I_1 \Delta V_1 = I_2 \Delta V_2$$

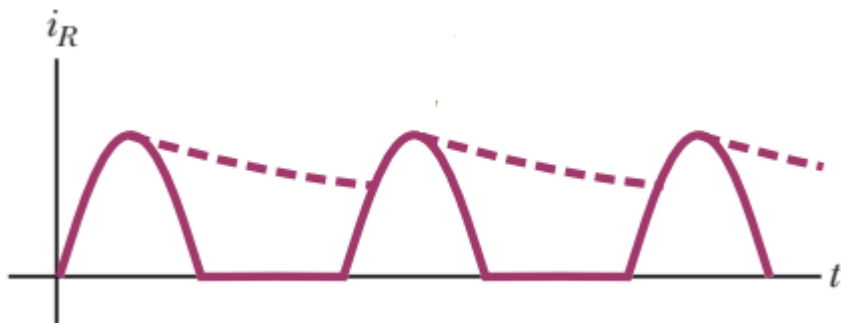
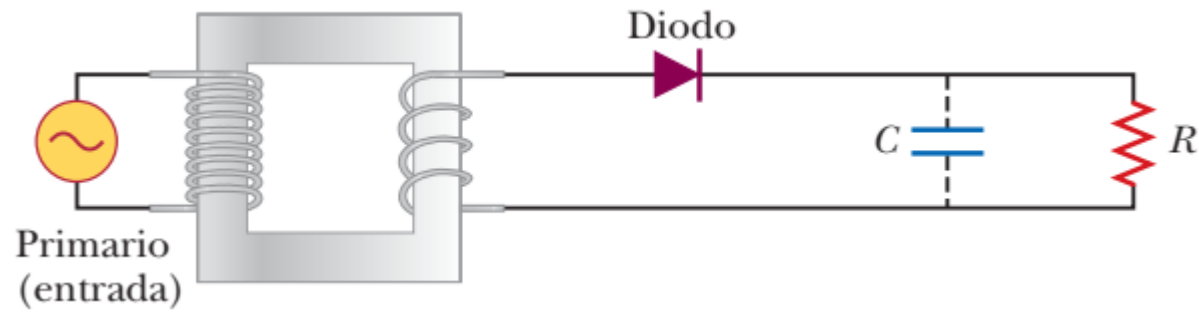


# Transformación y transmisión de potencia

$$R_{eq} = \left(\frac{N_1}{N_2}\right)^2 R_L$$



# Transformación y transmisión de potencia



# Instalaciones eléctricas domiciliarias

